

Fourier Analysis

(I) Fourier Analysis

- It is always possible to analyze „complex“ periodic waveforms into a set of sinusoidal waveforms
- Any periodic waveform can be approximated by adding together a number of sinusoidal waveforms
- Fourier analysis tells us what particular set of sinusoids go together to make up a particular complex waveform

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- The period is the duration of one cycle of an event and is the reciprocal of the frequency f . For example, if we count 40 events in two seconds, the frequency is

$$\frac{40}{2 \text{ s}} = \frac{20}{1 \text{ s}} = 20 \frac{1}{\text{s}} = 20 \text{ hertz}$$

- period is

$$T = p = \frac{1}{20} \text{ s.}$$

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- The frequency f is the inverse of the period

$$f = \frac{1}{T} = \frac{1}{p}$$

- If something changes rapidly, then we say that it has a high frequency.
- If it does not change rapidly, i.e., it changes smoothly, we say that it has a low frequency.

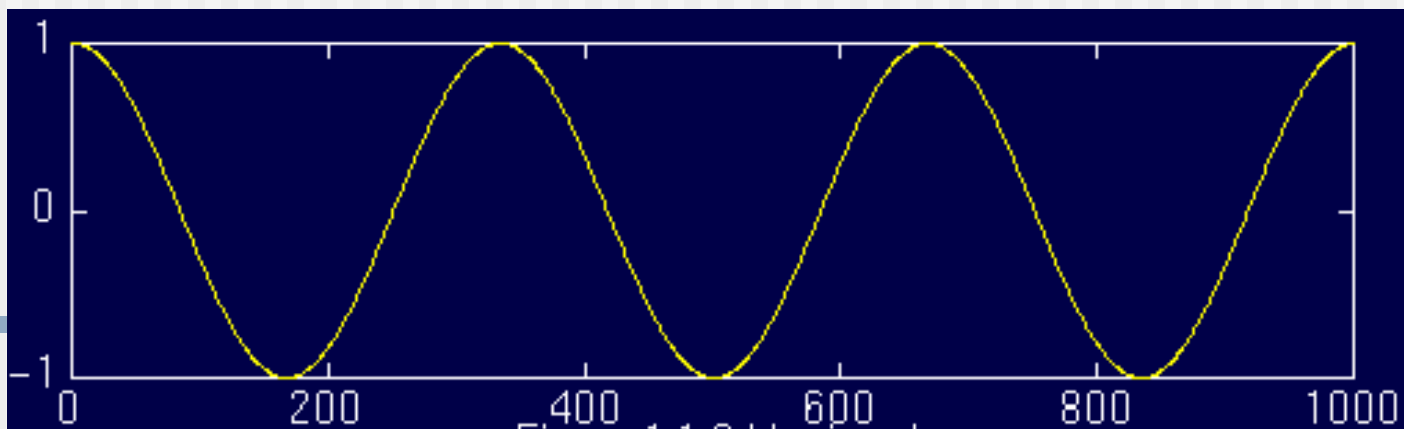


Figure 1.1 3 Hz signal

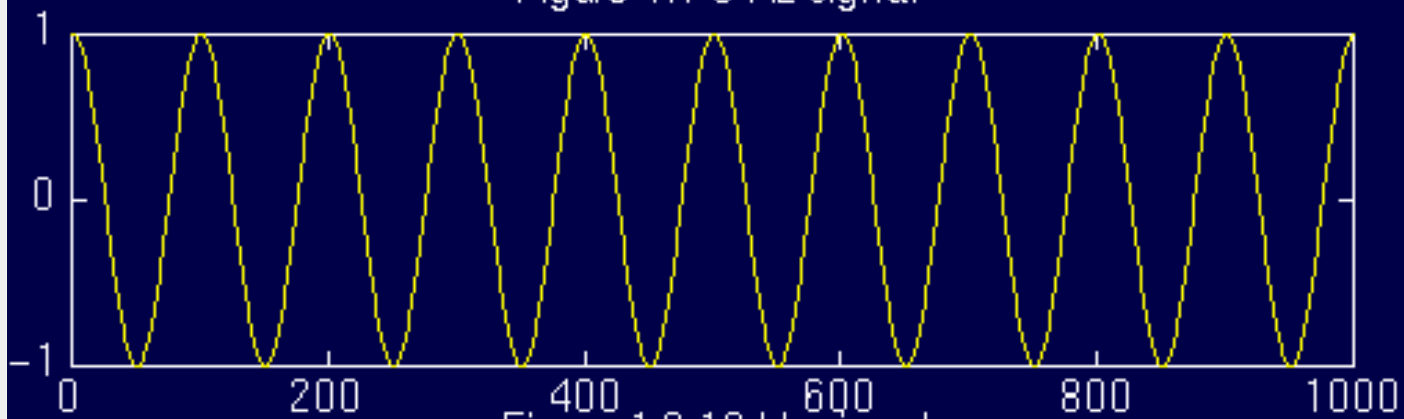


Figure 1.2 10 Hz signal

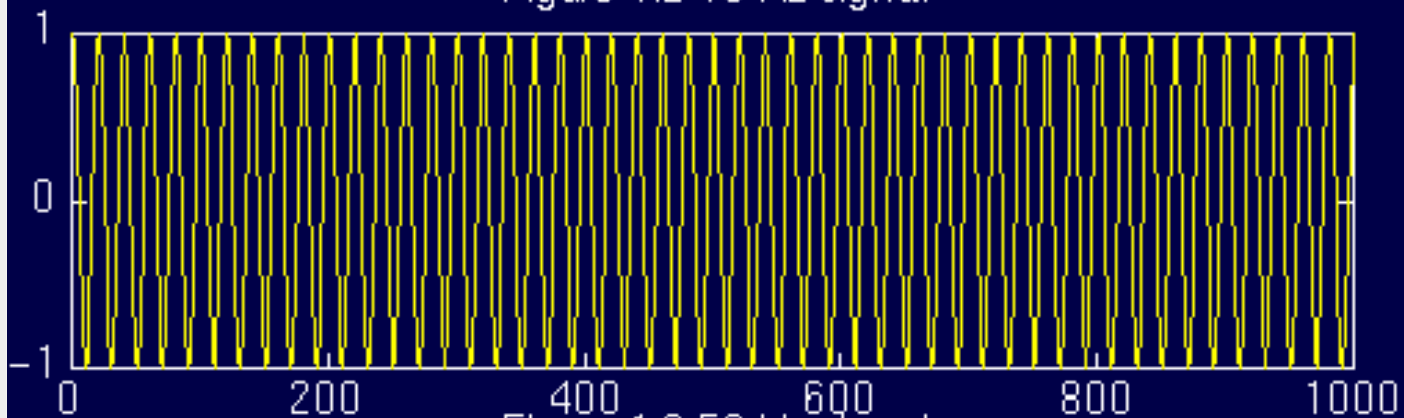
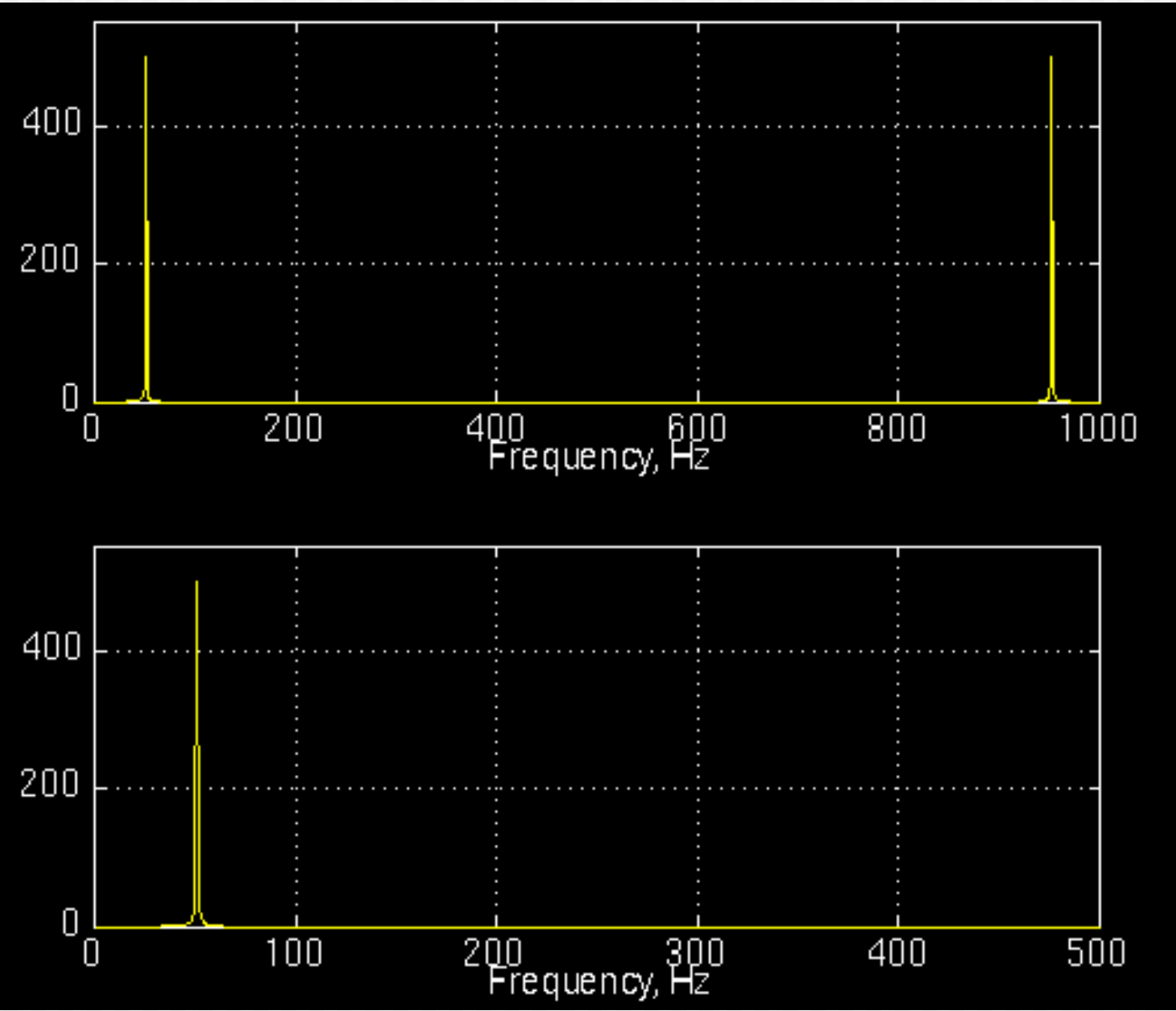


Figure 1.3 50 Hz signal

FOURIER TRANSFORM

- For example, if we take the FT of the electric current that we use in our houses,
- We will have one spike at 50 Hz
- Nothing elsewhere, since that signal has only 50 Hz frequency component



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- The frequency spectrum of a real valued signal is always symmetric. The top plot illustrates this point
 - However, since the symmetric part is exactly a mirror image of the first part
 - This symmetric second part is usually not shown

the Fourier transform of $x(t)$

$$X(f) = \int_{-\infty}^{\infty} x(t) \cdot e^{-2\pi i t f} dt$$

the inverse Fourier transform of $X(f)$

$$x(t) = \int_{-\infty}^{\infty} X(f) \cdot e^{2\pi i t f} df$$

- t stands for time, f stands for frequency, and x denotes the signal
- x denotes the signal in time domain and the X denotes the signal in frequency domain
- The signal $x(t)$, is multiplied with an exponential term, at some certain frequency " f ", and then integrated over **ALL TIMES** !

Discrete Fourier Transform

- Operates on discrete complex-valued function

- Given a function a :

$$a : [0, 1, \dots, N - 1] \rightarrow \mathbb{C}$$

- The discrete Fourier transform produces a function A :

$$A : [0, 1, \dots, N - 1] \rightarrow \mathbb{C}$$

$$A(x) = \frac{1}{\sqrt{N}} \sum_{k=0}^{N-1} a(k) \cdot e^{2\pi i \cdot \frac{kx}{N}}$$

- DFT can be seen as a linear transform taking the column vector \mathbf{a} to a column vector \mathbf{A}

$$\begin{pmatrix} A(0) \\ A(1) \\ \vdots \\ A(N-1) \end{pmatrix} = \begin{pmatrix} \frac{1}{\sqrt{N}} \cdot e^{2\pi i \frac{0 \cdot 0}{N}} & \frac{1}{\sqrt{N}} \cdot e^{2\pi i \frac{0 \cdot 1}{N}} & \dots & \frac{1}{\sqrt{N}} \cdot e^{2\pi i \frac{0 \cdot (N-1)}{N}} \\ \frac{1}{\sqrt{N}} \cdot e^{2\pi i \frac{1 \cdot 0}{N}} & \frac{1}{\sqrt{N}} \cdot e^{2\pi i \frac{1 \cdot 1}{N}} & \dots & \frac{1}{\sqrt{N}} \cdot e^{2\pi i \frac{1 \cdot (N-1)}{N}} \\ \vdots & \vdots & \ddots & \vdots \\ \frac{1}{\sqrt{N}} \cdot e^{2\pi i \frac{(N-1) \cdot 0}{N}} & \frac{1}{\sqrt{N}} \cdot e^{2\pi i \frac{(N-1) \cdot 1}{N}} & \dots & \frac{1}{\sqrt{N}} \cdot e^{2\pi i \frac{(N-1) \cdot (N-1)}{N}} \end{pmatrix} \cdot \begin{pmatrix} a(0) \\ a(1) \\ \vdots \\ a(N-1) \end{pmatrix}$$

■ Simplification

$$\begin{pmatrix} A(0) \\ A(1) \\ \vdots \\ A(N-1) \end{pmatrix} = \frac{1}{\sqrt{N}} \cdot \begin{pmatrix} 1 & 1 & \dots & 1 \\ 1 & e^{2\pi i \frac{1 \cdot 1}{N}} & \dots & e^{2\pi i \frac{(N-1) \cdot 1}{N}} \\ \vdots & \vdots & \ddots & \vdots \\ 1 & e^{2\pi i \frac{1 \cdot (N-1)}{N}} & \dots & e^{2\pi i \frac{(N-1) \cdot (N-1)}{N}} \end{pmatrix} \cdot \begin{pmatrix} a(0) \\ a(1) \\ \vdots \\ a(N-1) \end{pmatrix}$$

■ Example, N=4

$$\begin{pmatrix} A(0) \\ A(1) \\ \vdots \\ A(N-1) \end{pmatrix} = \frac{1}{2} \cdot \begin{pmatrix} 1 & 1 & 1 & 1 \\ 1 & i & -1 & -i \\ 1 & -1 & 1 & -1 \\ 1 & -i & -1 & i \end{pmatrix} \cdot \begin{pmatrix} a(0) \\ a(1) \\ \vdots \\ a(N-1) \end{pmatrix}$$

- Let $a : [0, 1, \dots, N - 1] \rightarrow C$ be a **periodic function**

$$a(x) = e^{-2\pi i \frac{ux}{N}}$$

$$a(x) = \cos\left(2\pi \frac{ux}{N}\right) + i \cdot \sin\left(2\pi \frac{ux}{N}\right)$$

$$e^{iu} = \cos(u) + i \cdot \sin(u)$$

```
In[30]:= TrigToExp[Cos[u] + i Sin[u]]
          e^{i u}

In[49]:= ExpToTrig[Exp[-2 * pi * i * u * x / N]]
Out[49]= Cos[2 pi u (pi : 0.1 : pi) / N] + i Sin[2 pi u (pi : 0.1 : pi) / N]

In[50]:=
          ExpToTrig[Exp[2 * pi * i * u * x / N]]
Out[50]= Cos[2 pi u (pi : 0.1 : pi) / N] - i Sin[2 pi u (pi : 0.1 : pi) / N]
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-
- A complex root of unity is a complex number $\omega^N = 1$
 - There are exactly n th roots of unity:

$$e^{2\pi i \frac{k}{N}} \quad \text{for } k = 0, 1, \dots, N - 1$$

- We define $\omega_N = e^{2\pi i \frac{1}{N}}$

$$e^{iu} = \cos(u) + i \cdot \sin(u)$$

$$\begin{pmatrix} A(0) \\ A(1) \\ \vdots \\ A(N-1) \end{pmatrix} = \frac{1}{\sqrt{N}} \cdot \begin{pmatrix} 1 & 1 & \dots & 1 \\ 1 & e^{2\pi i \frac{1 \cdot 1}{N}} & \dots & e^{2\pi i \frac{(N-1) \cdot 1}{N}} \\ \vdots & \vdots & \ddots & \vdots \\ 1 & e^{2\pi i \frac{1 \cdot (N-1)}{N}} & \dots & e^{2\pi i \frac{(N-1) \cdot (N-1)}{N}} \end{pmatrix} \cdot \begin{pmatrix} a(0) \\ a(1) \\ \vdots \\ a(N-1) \end{pmatrix}$$

$$\begin{pmatrix} A(0) \\ A(1) \\ \vdots \\ A(N-1) \end{pmatrix} = \frac{1}{\sqrt{N}} \cdot \begin{pmatrix} 1 & 1 & \dots & 1 \\ 1 & \omega_N^{1 \cdot 1} & \dots & \omega_N^{(N-1) \cdot 1} \\ \vdots & \vdots & \ddots & \vdots \\ 1 & \omega_N^{1 \cdot (N-1)} & \dots & \omega_N^{(N-1) \cdot (N-1)} \end{pmatrix} \cdot \begin{pmatrix} a(0) \\ a(1) \\ \vdots \\ a(N-1) \end{pmatrix}$$

Remarks

$$\begin{pmatrix} A(0) \\ A(1) \\ \vdots \\ A(N-1) \end{pmatrix} = \frac{1}{\sqrt{N}} \cdot \begin{pmatrix} 1 & 1 & \dots & 1 \\ 1 & \omega_N^{1 \cdot 1} & \dots & \omega_N^{(N-1) \cdot 1} \\ \vdots & \vdots & \ddots & \vdots \\ 1 & \omega_N^{1(N-1)} & \dots & \omega_N^{(N-1)(N-1)} \end{pmatrix} \cdot \begin{pmatrix} a(0) \\ a(1) \\ \vdots \\ a(N-1) \end{pmatrix}$$

$$\begin{pmatrix} y_0 \\ y_1 \\ \vdots \\ y_{N-1} \end{pmatrix} = \frac{1}{\sqrt{N}} \cdot \begin{pmatrix} 1 & 1 & \dots & 1 \\ 1 & \omega_N^{1 \cdot 1} & \dots & \omega_N^{(N-1) \cdot 1} \\ \vdots & \vdots & \ddots & \vdots \\ 1 & \omega_N^{1(N-1)} & \dots & \omega_N^{(N-1)(N-1)} \end{pmatrix} \cdot \begin{pmatrix} x_0 \\ x_1 \\ \vdots \\ x_{N-1} \end{pmatrix}$$

- Input vector of complex numbers of length N

$$x_0, x_1, \dots, x_{N-1}$$

$$y_0, y_1, \dots, y_{N-1}$$

$$y_k = \frac{1}{\sqrt{N}} \sum_{j=0}^{N-1} x_j e^{-\frac{2\pi i}{N}kj} \quad k \in \{0, 1, \dots, N-1\}$$

inverse :

$$x_k = \frac{1}{\sqrt{N}} \sum_{j=0}^{N-1} y_j e^{\frac{2\pi i}{N}kj} \quad j \in \{0, 1, \dots, N-1\}$$

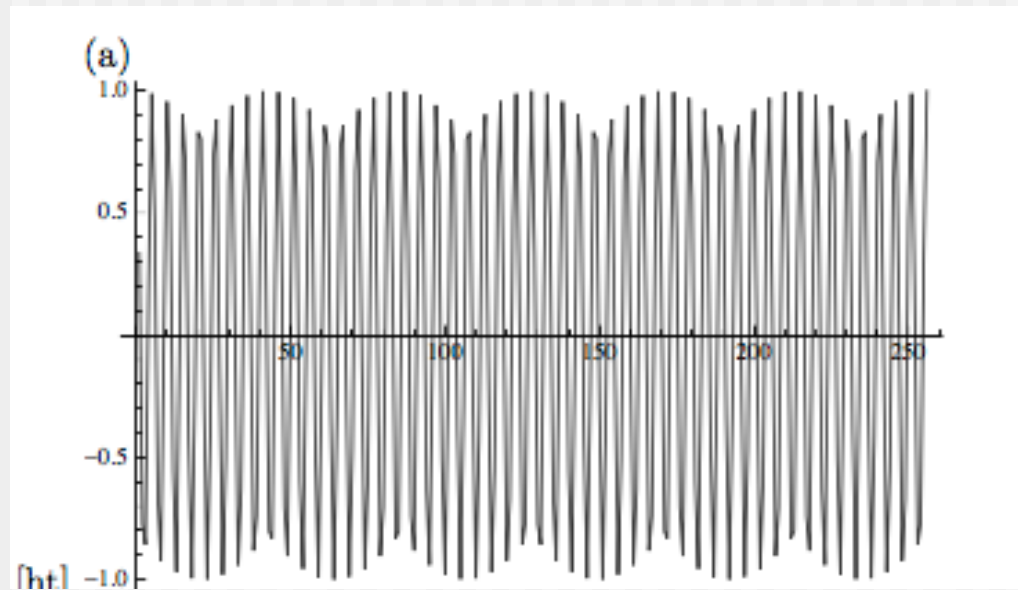
Noise reduction

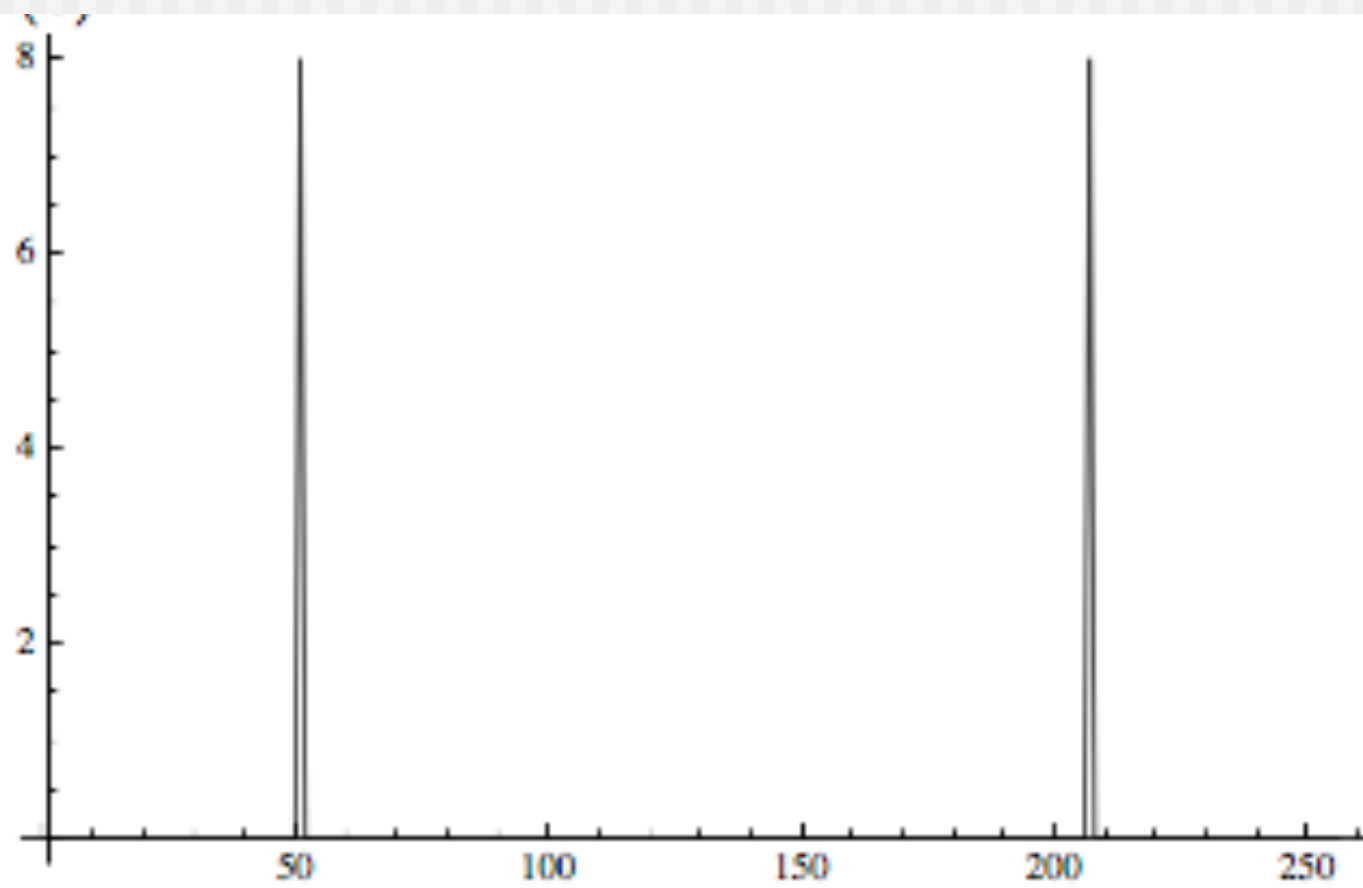
- It is difficult to identify the frequency components by looking at the original signal
- Converting to the frequency domain
- If dimension reduction, store only a fraction of frequencies (with high amplitude)
- If noise reduction
 - (remove high frequencies, fast change, smoothing)
 - (remove low frequencies, slow change, remove global trends)
 - Inverse discrete Fourier transform

Example

We generate a list with $256 = 2^8$ elements containing a periodic signal α_t

$$\alpha_t = \cos\left(\frac{50 \cdot t \cdot 2 \cdot \pi}{256}\right),$$



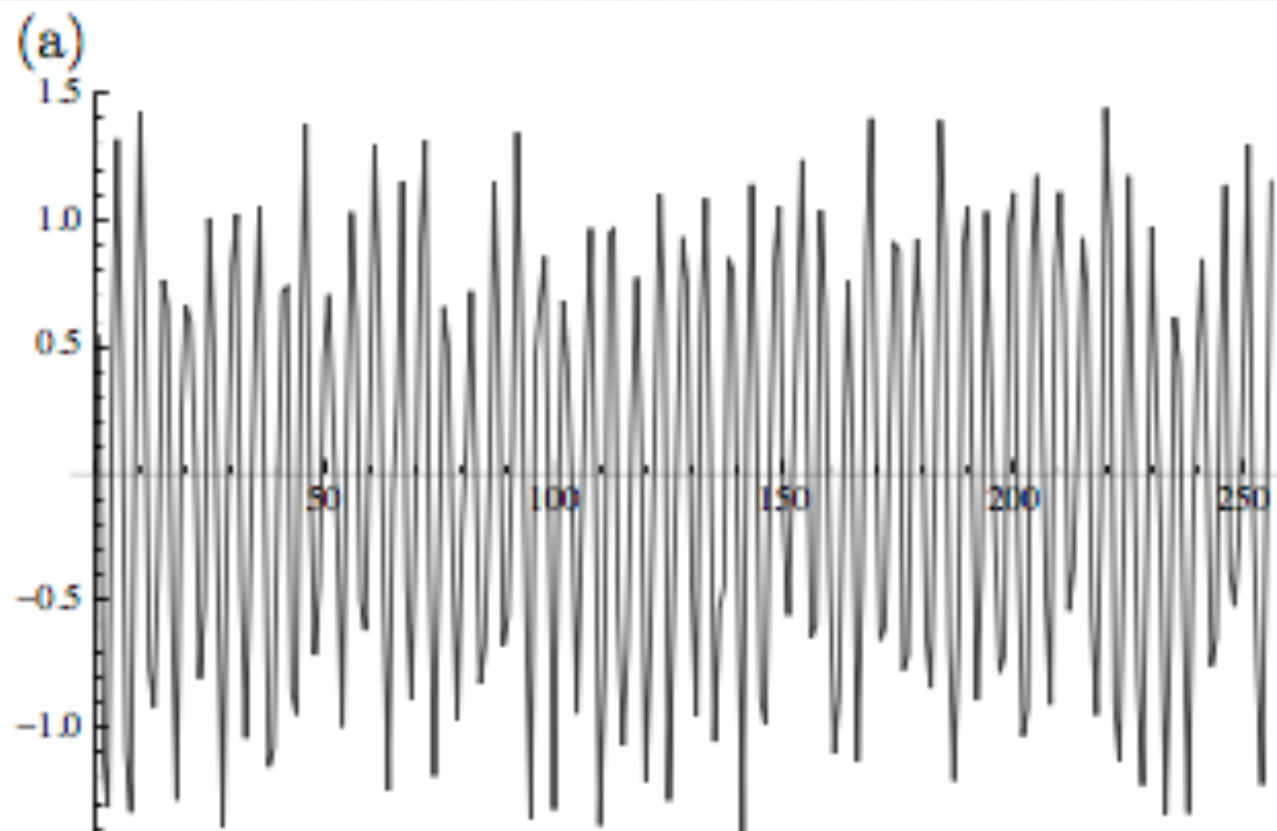


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- The discrete Fourier transform ω_f of the real valued signal $x[n]$ is symmetric. It shows a strong peak at $50 + 1$ and a symmetric peak at $256 - 50 + 1$ representing the frequency component of the signal

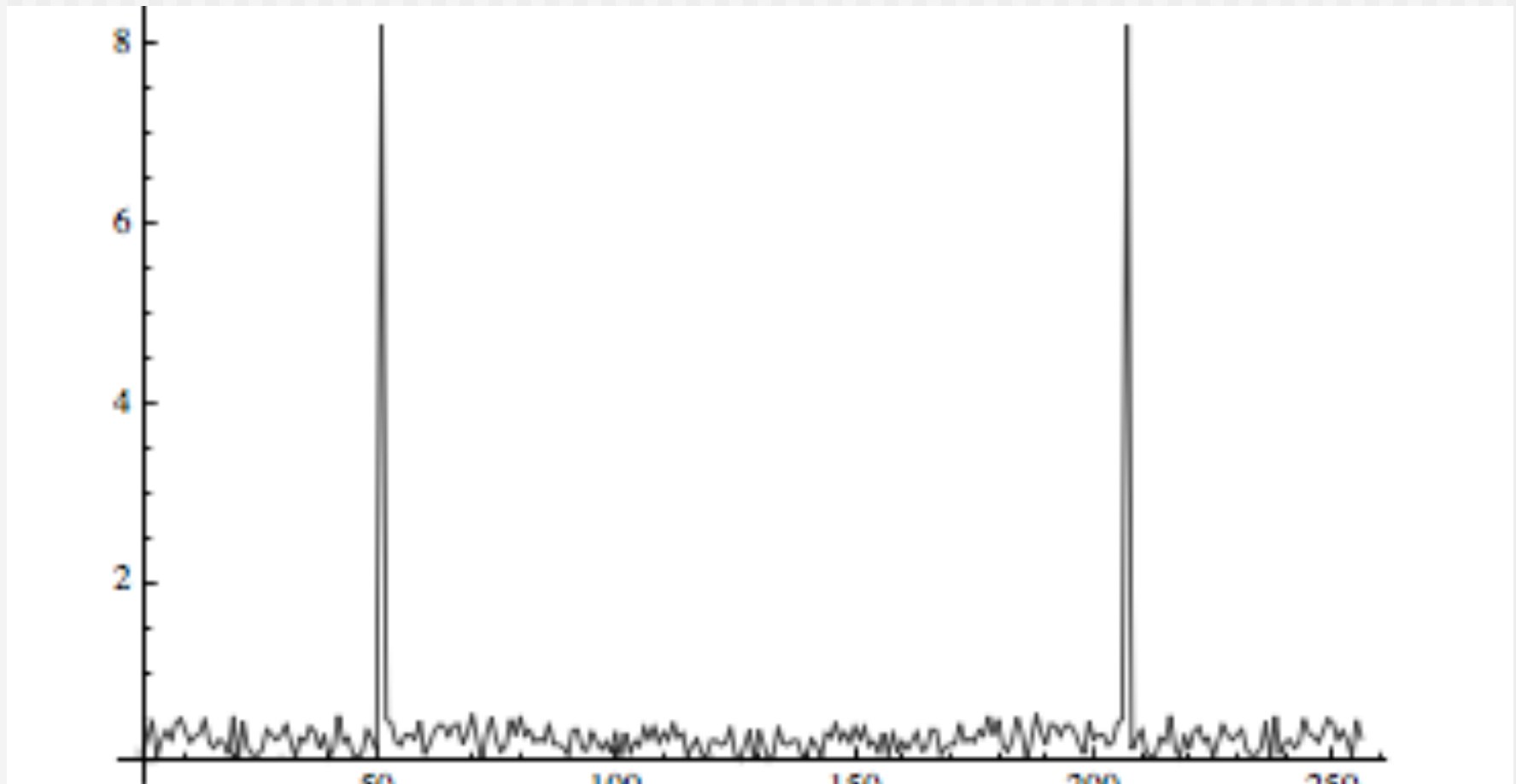
-
- We add to the periodic signal a Gaussian random noise from the interval $[-0.5, 0.5]$.

$$\alpha_t^* = \cos\left(\frac{50 \cdot t \cdot 2 \cdot \pi}{256}\right) + \text{noise.}$$

- The represented data looks random

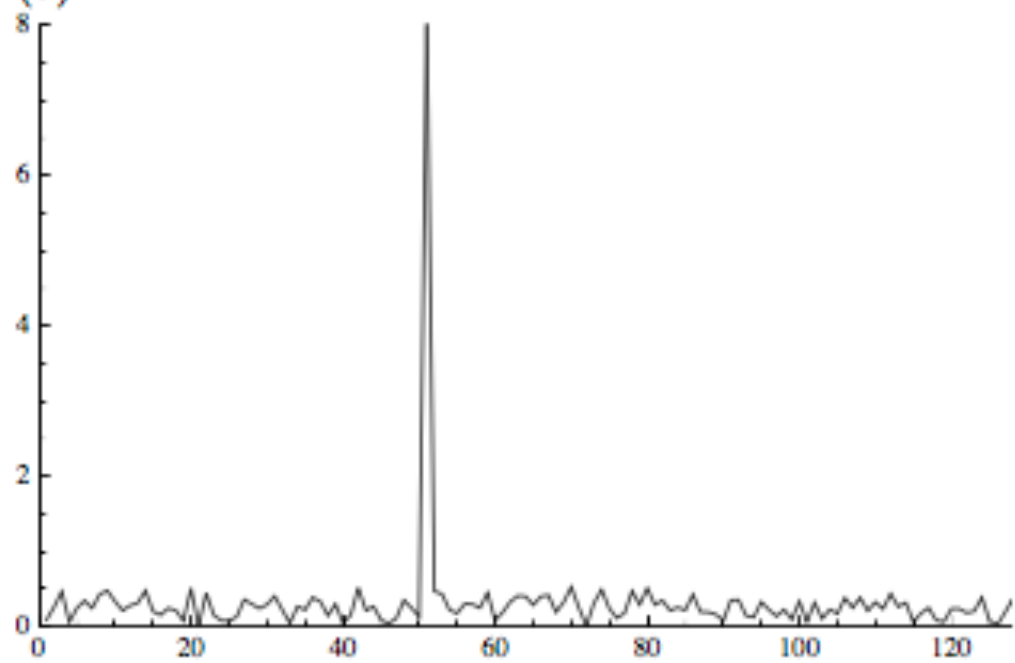


The frequency component

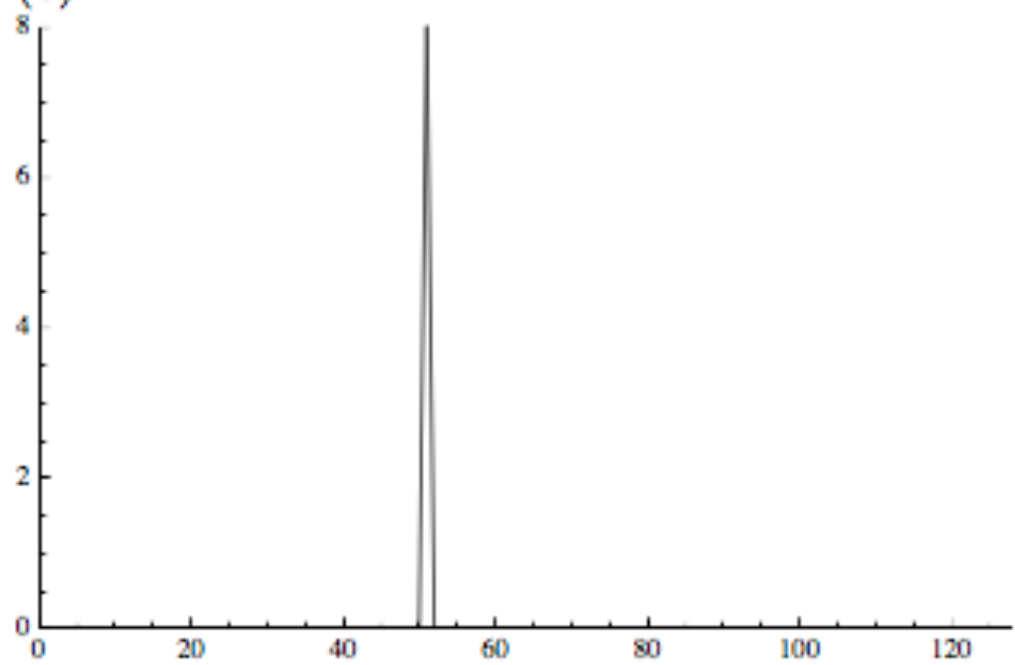


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- A filter that reduces Gaussian noise based on DFT removes frequencies with low amplitude of ω_f and performs the inverse discrete Fourier transform
 - For dimension reduction of the signal, only a fraction of frequencies with high amplitude are represented.

(a)



(b)



Feature space

- Sample $\{\vec{x}^{(1)}, \vec{x}^{(2)}, \dots, \vec{x}^{(k)}, \dots, \vec{x}^{(n)}\}$

$$\vec{x} = \begin{cases} x_1 \\ x_2 \\ \dots \\ \dots \\ x_d \end{cases} \in \mathfrak{R}^d \quad \|\vec{x} - \vec{y}\| = \sqrt{\sum_{i=1}^d (x_i - y_i)^2}$$

Scaling

- A well-known scaling method consists of performing some scaling operations
 - subtracting the mean and dividing the standard deviation

$$y_i = \frac{(x_i - m_i)}{s_i}$$

- m_i sample mean
- s_i sample standard deviation

- According to the scaled metric the **scaled** feature vector is expressed as

$$\|\vec{y}\|_s = \sqrt{\sum_{i=1}^n \frac{(x_i - m_i)^2}{s_i^2}}$$

- shrinking **large** variance values
 - $s_i > 1$
- stretching **low** variance values
 - $s_i < 1$
- Fails to preserve distances when *general linear transformation* is applied!

Covariance

- Covariance
 - Measuring the tendency two features x_i and x_j varying in the **same direction**
 - The covariance between features x_i and x_j is estimated for n patterns

$$C_{ij} = \frac{\sum_{k=1}^n (x_i^{(k)} - m_i)(x_j^{(k)} - m_j)}{n - 1}$$

$$C = \begin{bmatrix} c_{11} & c_{12} & \dots & c_{1d} \\ c_{21} & c_{22} & \dots & c_{2d} \\ \dots & \dots & \dots & \dots \\ c_{d1} & c_{d2} & \dots & c_{dd} \end{bmatrix}$$

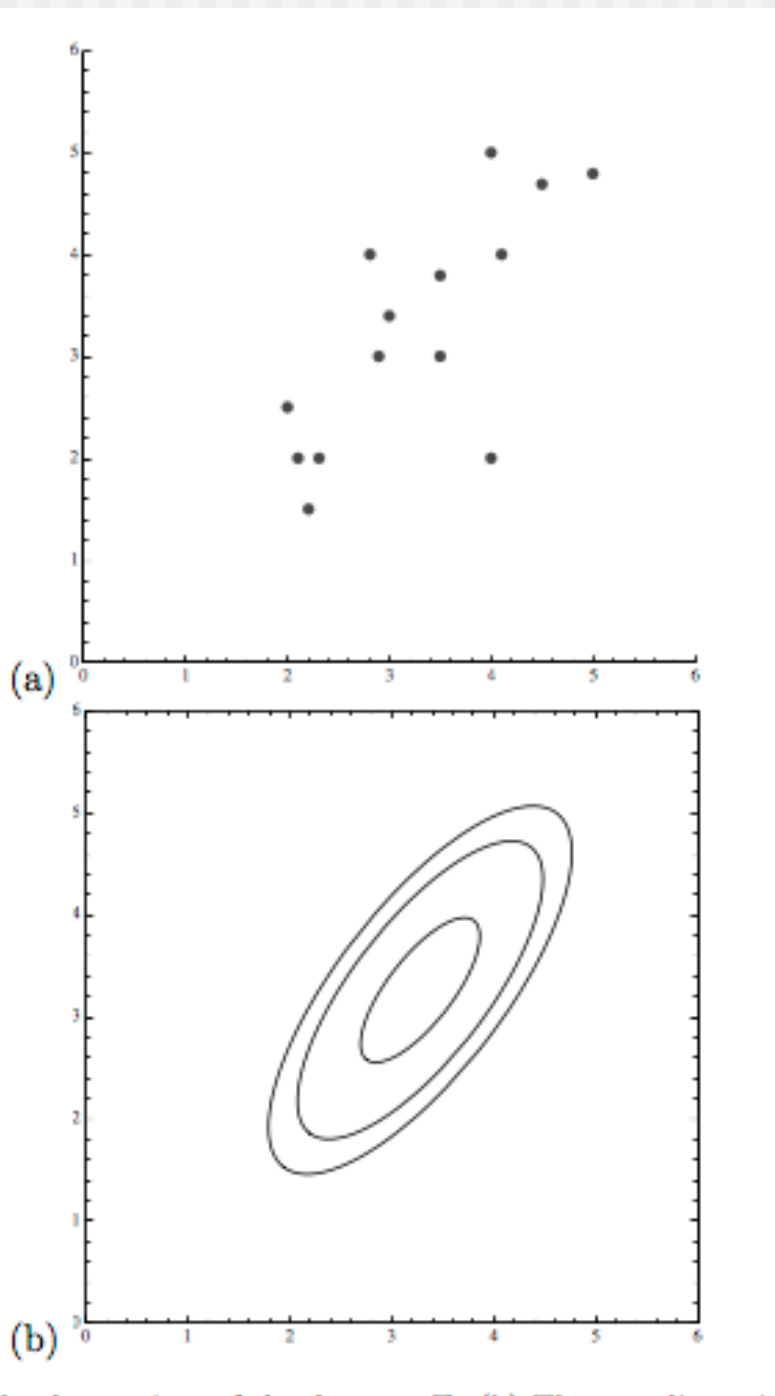
Correlation

- Covariances are symmetric $c_{ij}=c_{ji}$
- Covariance is related to correlation

$$r_{ij} = \frac{\sum_{k=1}^n (x_i^{(k)} - m_i)(x_j^{(k)} - m_j)}{(n-1)s_i s_j} = \frac{c_{ij}}{s_i s_j} \in [-1,1]$$

$$\Sigma = \{(2.1, 2), (2.3, 2), (2.9, 3), (4.1, 4), (5, 4.8), (2, 2.5), (2.2, 1.5), (4, 5), (4, 2), (2.8, 4), (3, 3.4), (3.5, 3.8), (4.5, 4.7), (3.5, 3)\}$$

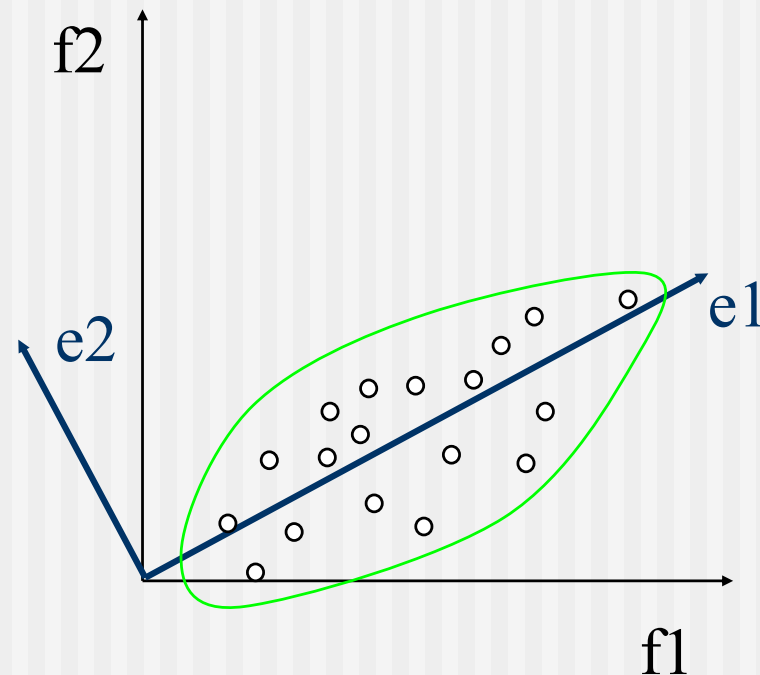
$$c_{ij} = \frac{\sum_{k=1}^n (x_{k,i} - \bar{x}_i) \cdot (y_{k,j} - \bar{y}_j)}{n - 1}$$



$$C = \begin{pmatrix} 0.912582 & 0.82456 \\ 0.82456 & 1.34247 \end{pmatrix}$$

Principal Component Analysis

- Intuition: find the axis that shows the greatest variation, and project all points into this axis



(II) Karhunen-Loève Transformation

- Covariance matrix C of (a $d \times d$ matrix)
 - Symmetric and positive definite

$$U^T C U = \Lambda = \text{diag}(\lambda_1, \lambda_2, \dots, \lambda_d)$$

$$(\lambda I - C)u = 0$$

- There are d eigenvalues and eigenvectors

$$C \vec{u}_i = \lambda_i \vec{u}_i$$

- is the λ_i i th eigenvalue of C and \vec{u}_i the i th column of \mathbf{U} , the i th eigenvectors

-
- Eigenvectors are always **orthogonal**
 - U is an orthonormal matrix $UU^T=U^TU=I$
 - U defines the K-L transformation
 - The transformed features by the K-L transformation are given by

$$\vec{y} = U^T \vec{x} \quad (\text{linear Transformation})$$

- K-L transformation rotates the feature space into alignment with **uncorrelated** features

Example

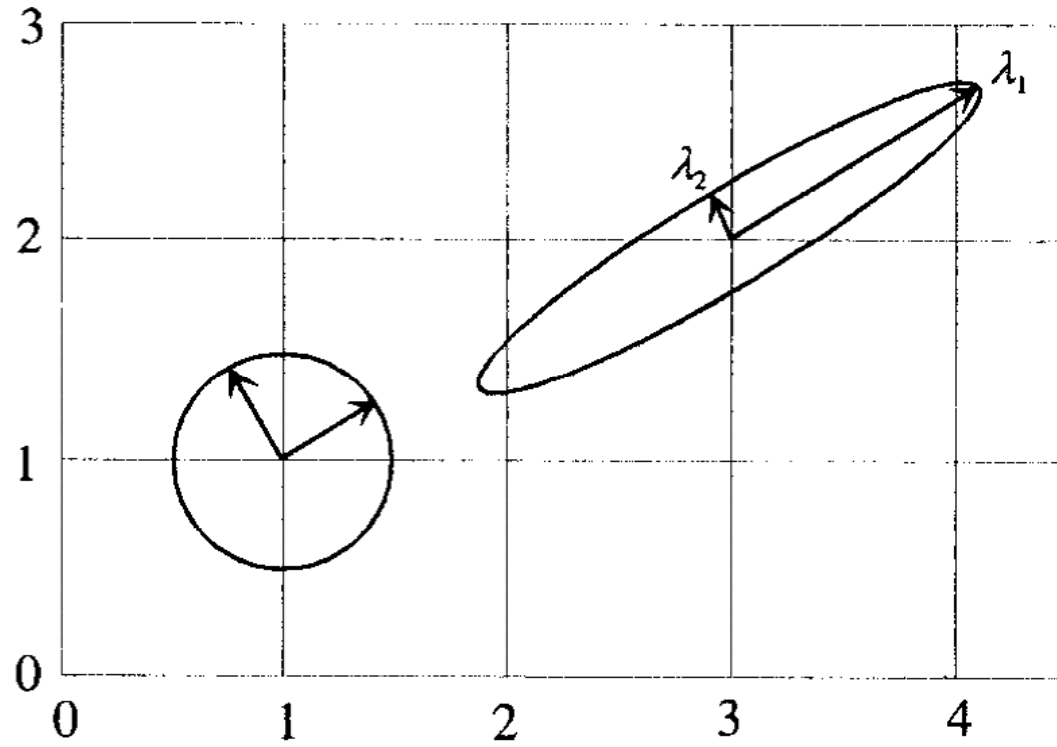
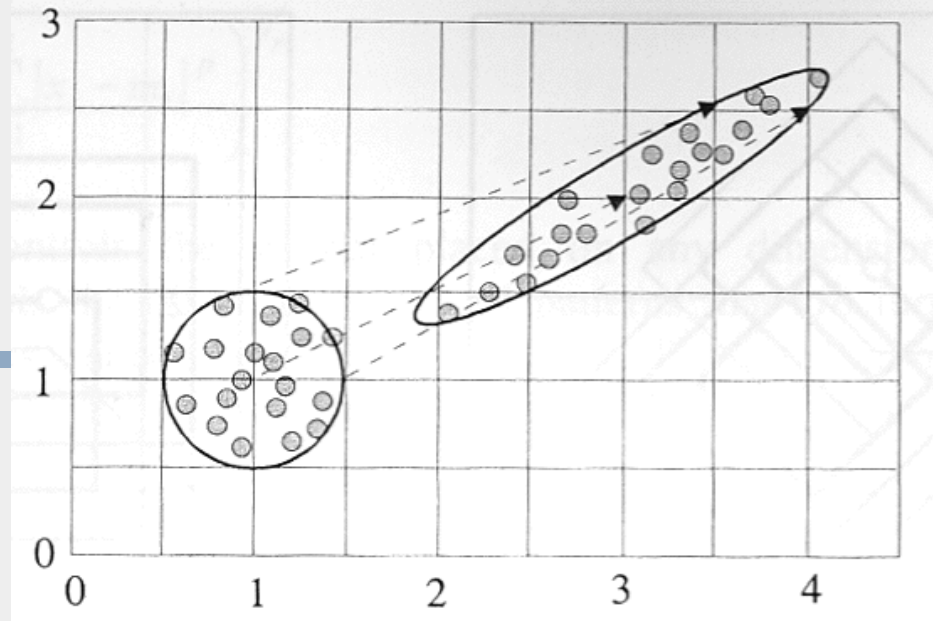
$$C = \begin{bmatrix} 2 & 1 \\ 1 & 1 \end{bmatrix} \quad |\lambda I - C| = 0 \quad \lambda^2 - 3\lambda + 1 = 0$$

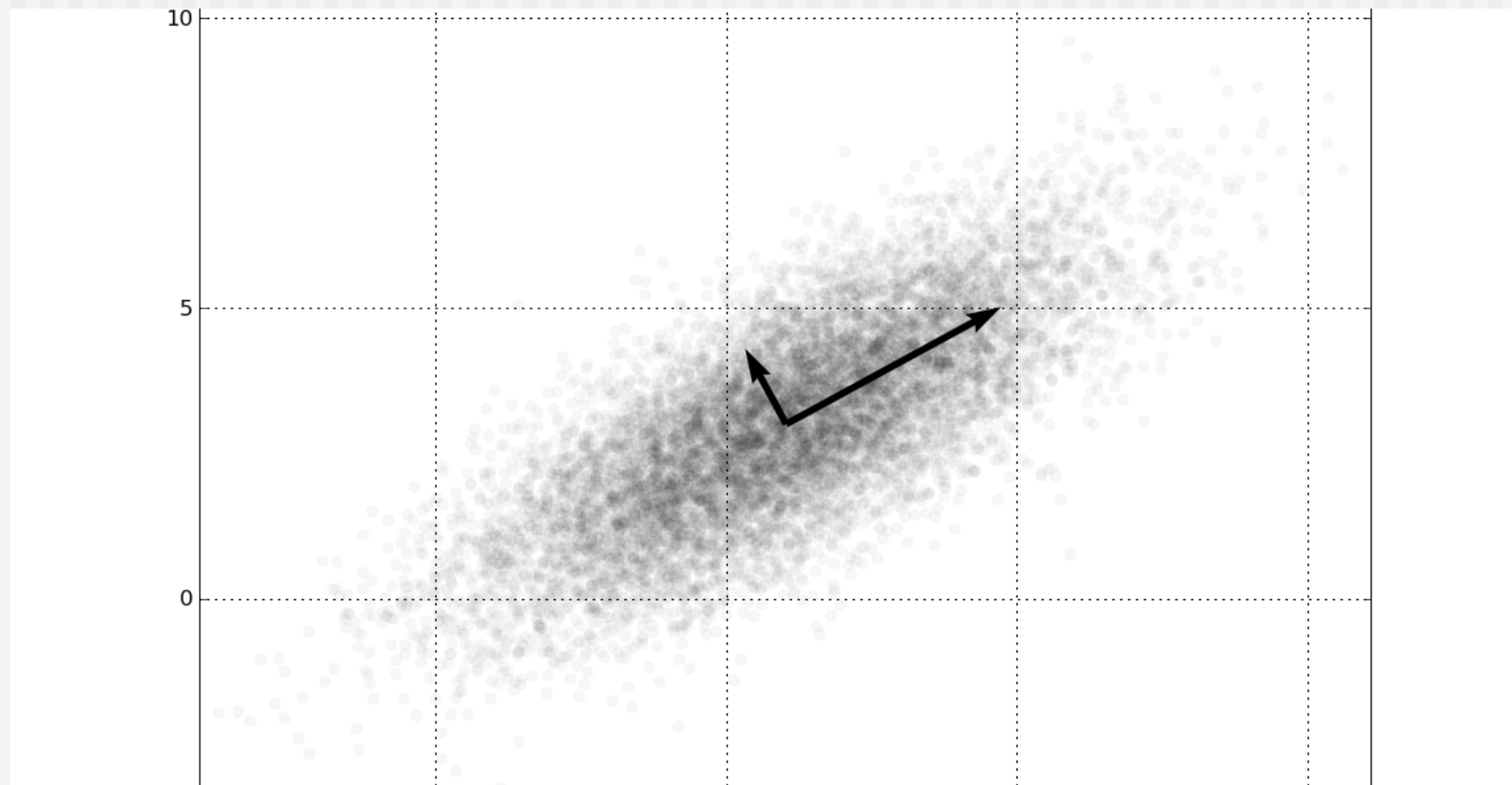
- $\lambda_1 = 2.618 \quad \lambda_2 = 0.382$

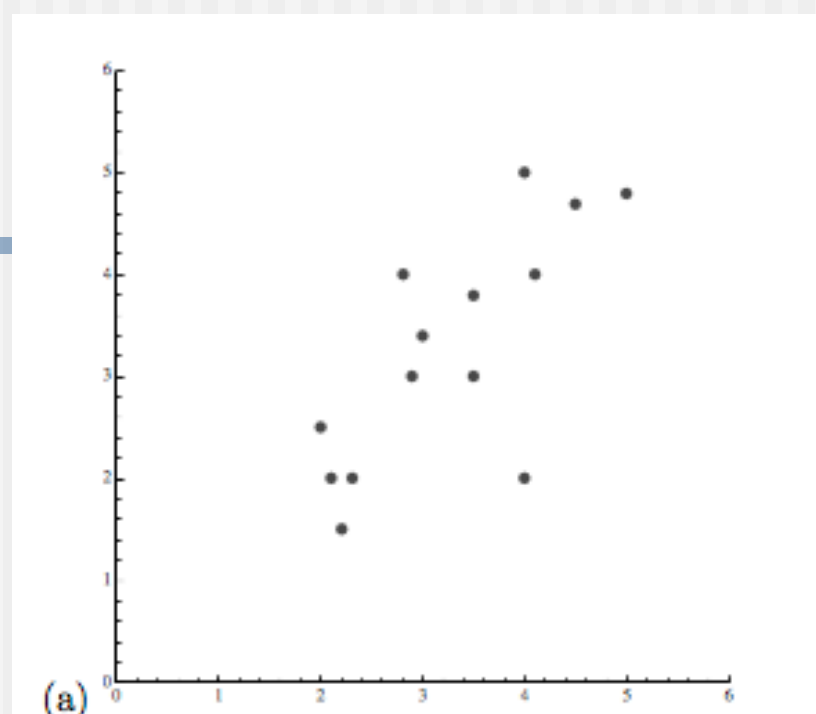
$$\begin{bmatrix} -0.618 & -1 \\ -1 & 1.618 \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \end{bmatrix} = 0$$

- $u^{(1)} = [1 \ 0.618] \quad u^{(2)} = [-1 \ 1.618]$

- $\mathbf{U} = [u^{(1)}, u^{(2)}]$







The squares of the eigenvalues represent the variances along the eigenvectors. The eigenvalues corresponding to the covariance matrix of the data set Σ are

$$\lambda_1 = 1.97964, \quad \lambda_2 = 0.275412$$

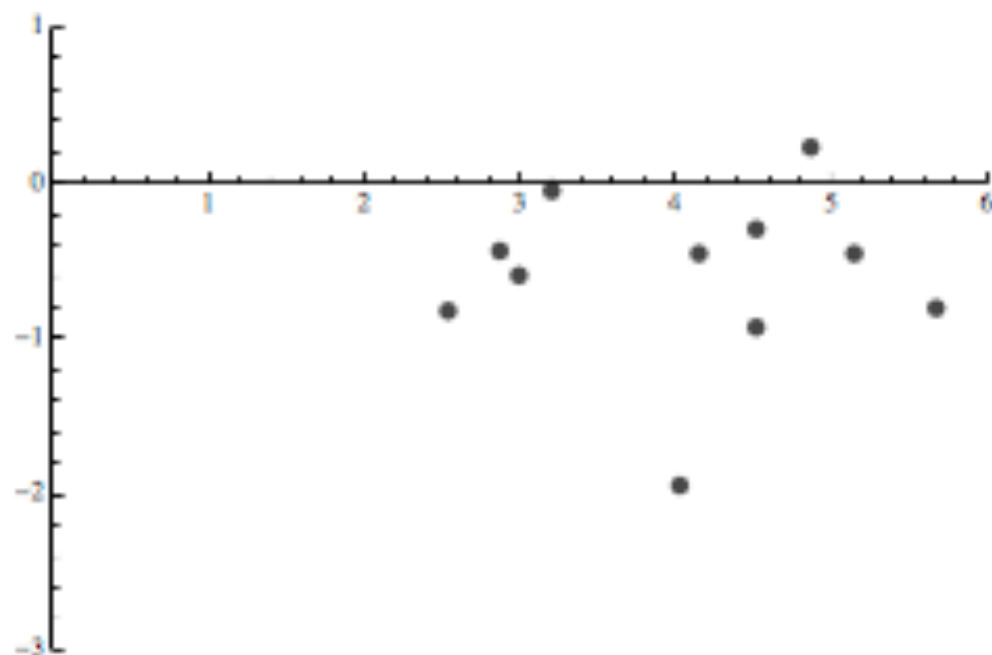
and the corresponding normalized eigenvectors are

$$\mathbf{u}_1 = \begin{pmatrix} 0.611454 \\ 0.79128 \end{pmatrix}, \quad \mathbf{u}_2 = \begin{pmatrix} -0.79128 \\ 0.611454 \end{pmatrix}.$$

Plot the data set with the matrix U with

$$U = \begin{pmatrix} 0.611454 & -0.79128 \\ 0.79128 & 0.611454 \end{pmatrix}.$$

$$\mathbf{y} = U^T \cdot \mathbf{x} = \begin{pmatrix} 0.611454 & 0.79128 \\ -0.79128 & 0.611454 \end{pmatrix} \cdot \mathbf{x},$$



PCA (Principal Components Analysis)

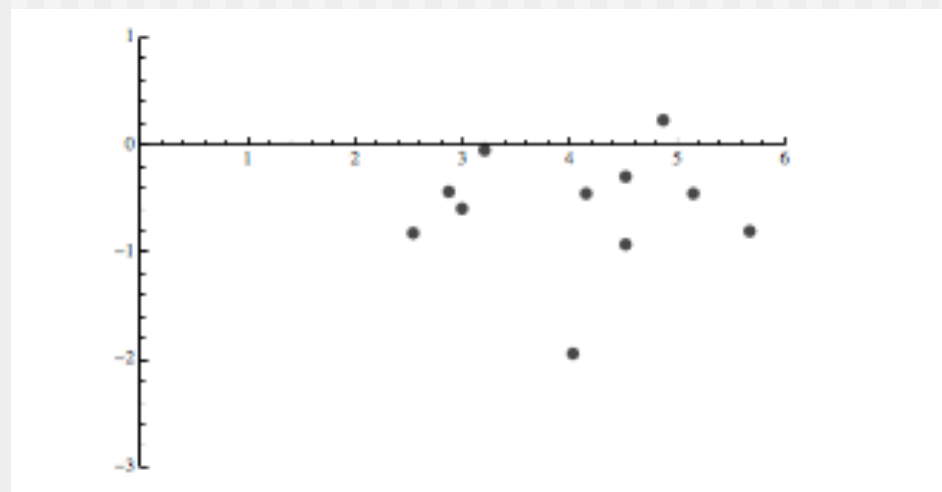
- New features \mathbf{y} are uncorrelated with the covariance Matrix
- Each eigenvector \mathbf{u}_i is **associated** with some variance associated by λ_i
- Uncorrelated features with **higher variance** (represented by λ_i) contain **more information**
- Idea:
 - Retain only the significant eigenvectors \mathbf{u}_i
 - Example
 - $\mathbf{U}=[\mathbf{u}^{(1)},\mathbf{u}^{(2)}]$ $\lambda_1=2.618$ $\lambda_2=0.382$
 - $\mathbf{U}^*=[\mathbf{u}^{(1)}]$
 -

$$\vec{y} = \mathbf{U}^{*T} \vec{x}$$

Dimension Reduction

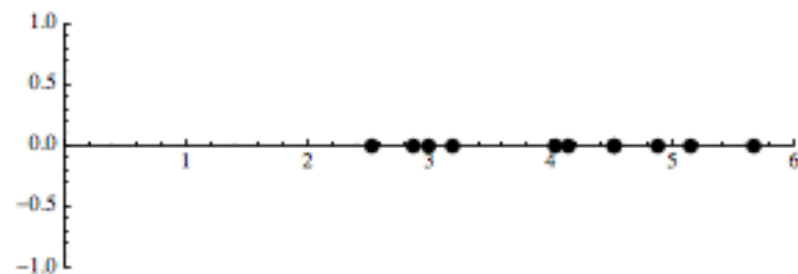
- How many eigenvectors (and corresponding eigenvalue) to retain
- Kaiser criterion
 - Discards eigenvectors whose eigenvalues are below 1

$$\mathbf{y} = U^T \cdot \mathbf{x} = \begin{pmatrix} 0.611454 & 0.79128 \\ -0.79128 & 0.611454 \end{pmatrix} \cdot \mathbf{x},$$



$$\mathbf{z} = W^T \cdot \mathbf{x}$$

$$\mathbf{z} = W^T \cdot \mathbf{x} = (0.611454 \ 0.79128) \cdot \mathbf{x}$$



Problems

- Principal components are linear transformation of the original features
- It is difficult to attach any semantic meaning to principal components
- For new data which is added to the dataset, the PCA has to be recomputed

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- Suppose we have a covariance matrix:

$$C = \begin{pmatrix} 3 & 1 \\ 1 & 21 \end{pmatrix}$$

- What is the matrix of the K-L transformation?

-
- First we have to compute the eigenvalues
 - The system has to become linear dependent (singular)

$$|\lambda I - C| = 0$$

- The determinant has to become zero

$$\lambda^2 - 24\lambda + 62 = 0$$

■ Solving it we get

■ $\lambda_1 = 2,94461$

■ $\lambda_2 = 21,05538$

Now, lets compute the two eigenvectors....

- To do it you have to solve two singular, dependent systems
- for the first eigenvalue $\lambda_1=2,94461$
 $(\lambda_1 I - C)\vec{u}_1 = 0$
- Or if you prefer more ...

$$\lambda_1 \vec{u}_1 = C \vec{u}_1$$

$$C \vec{u}_1 = \lambda_1 \vec{u}_1$$

Now, lets compute the two eigenvectors....

- To do it you have to solve two singular, dependent systems
- And for the second eigenvalue $\lambda_2=21,05538$
 $(\lambda_2 I - C)\vec{u}_2 = 0$
- Or if you prefer more ...

$$\lambda_2 \vec{u}_2 = C \vec{u}_2$$

$$C \vec{u}_2 = \lambda_2 \vec{u}_2$$

For $\lambda_1=2,94461$

$\mathbf{u}_1=(u_1,u_2)$

$$\left[\begin{array}{cc} 2,94461 & 0 \\ 0 & 2,94461 \end{array} \right] - \left[\begin{array}{cc} 3 & 1 \\ 1 & 21 \end{array} \right] \begin{bmatrix} u_1 \\ u_2 \end{bmatrix} = 0$$

■ We have to find a nontrivial solution!

- Trivial solution is $u=[0,0]$...

$$\left[\begin{array}{cc} -0,05539 & -1 \\ -1 & -18,055 \end{array} \right] \begin{bmatrix} u_1 \\ u_2 \end{bmatrix} = 0$$

-
- Because the system is linear dependable, the left column is multiple value of the right column
 - There are infinity many solution!!!!

 - We have only to determine the direction of the eigenvectors \mathbf{u}_1 and \mathbf{u}_2
 - But be careful, the *normalized* vectors have to be orthogonal to each other
 - $\langle \mathbf{u}_1, \mathbf{u}_2 \rangle = 0$

- Let be $u_1=1$ then we have to determine u_2

$$\begin{bmatrix} -0,05539 & -1 \\ -1 & -18,055 \end{bmatrix} \begin{bmatrix} 1 \\ u_2 \end{bmatrix} = 0$$

$$\begin{bmatrix} -0,05539 \\ -1 \end{bmatrix} = \begin{bmatrix} 1 \\ 18,055 \end{bmatrix} u_2$$

- $\mathbf{u}_1 = [u_1, u_2] = [1, -0,05539]$

For $\lambda_2 = 21,05538$

$\mathbf{u}_1 = (u_1, u_2)$

$$\left[\begin{bmatrix} 21,05538 & 0 \\ 0 & 21,05538 \end{bmatrix} - \begin{bmatrix} 3 & 1 \\ 1 & 21 \end{bmatrix} \right] \begin{bmatrix} u_1 \\ u_2 \end{bmatrix} = 0$$

- We have to find a nontrivial solution!
 - Trivial solution is $u = [0, 0]$... Déjà vu?

$$\begin{bmatrix} 18,055 & -1 \\ -1 & 0,05538 \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \end{bmatrix} = 0$$

-
- Let be $u_1=1$ then we have to determine u_2

$$\begin{bmatrix} 18,055 \\ -1 \end{bmatrix} = \begin{bmatrix} 1 \\ -0,05538 \end{bmatrix} u_2$$

- $\mathbf{u}_2 = [u_1, u_2] = [1, 18,055]$

- $\mathbf{u}_1 = [u_1, u_2] = [1, -0,05539]$

- for $\lambda_1 = 2,94461$

- $\mathbf{u}_2 = [u_1, u_2] = [1, 18,055]$

- $\lambda_2 = 21,05538$

- Orthogonal? Yes $\langle \mathbf{u}_1, \mathbf{u}_2 \rangle = 0$

- Which of the two eigenvectors is more significant?

- \mathbf{u}_2 , because $\lambda_1 = 2,94461 < \lambda_2 = 21,05538$

- Remember, we have to normalize the Eigenvectors